

# **MTI Noise Integration Loss**

**G. V. TRUNK**

*Radar Analysis Staff  
Radar Division*

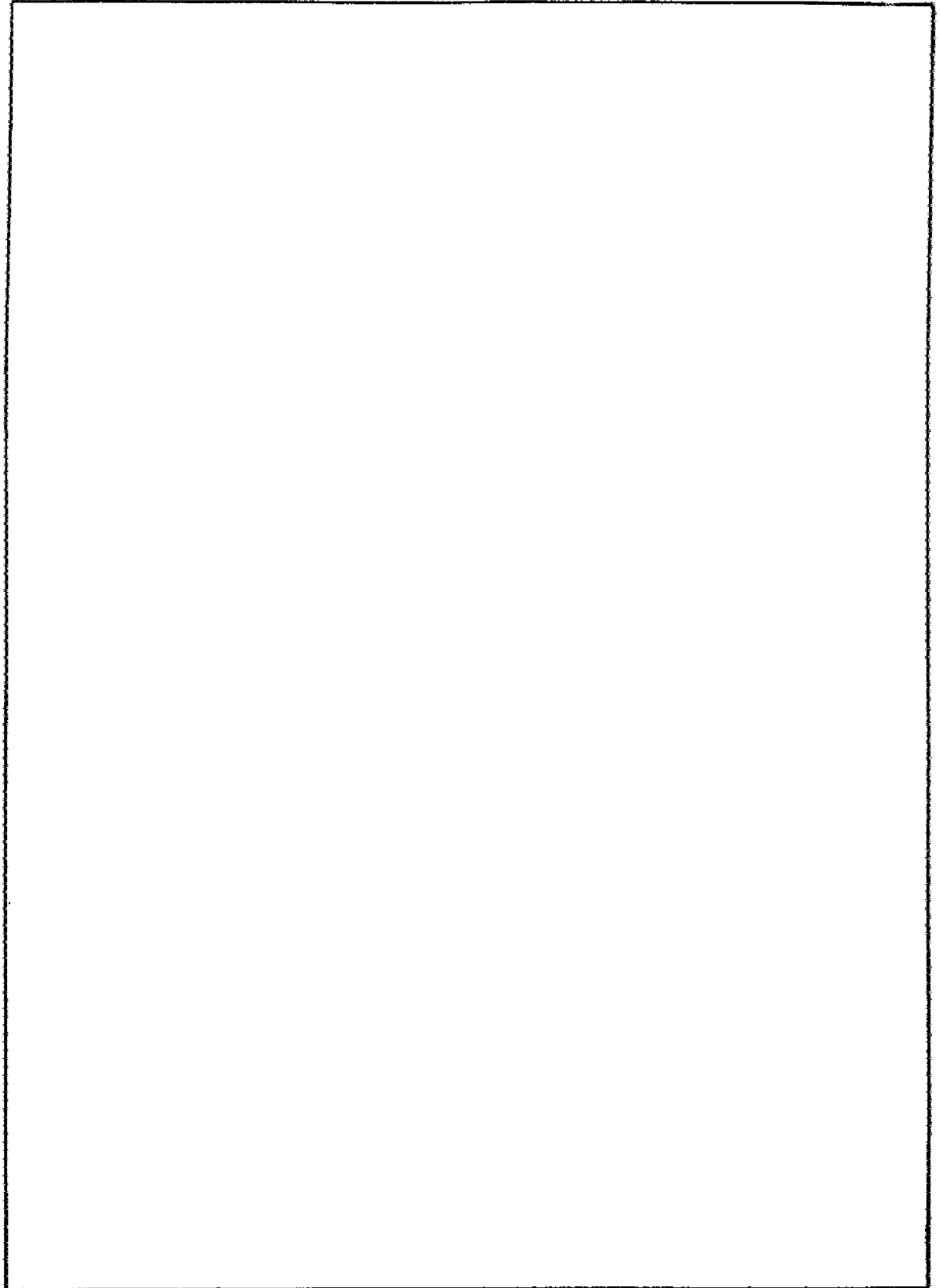
July 15, 1977



**NAVAL RESEARCH LABORATORY  
Washington, D.C.**

Approved for public release; distribution unlimited.

REPORT DOCUMENTATION PAGE		READ INSTRUCTIONS BEFORE COMPLETING FORM
1. REPORT NUMBER NRL Report 8132	2. GOVT ACCESSION NO.	3. RECIPIENT'S CATALOG NUMBER
4. TITLE (and Subtitle) MTI NOISE INTEGRATION LOSS		5. TYPE OF REPORT & PERIOD COVERED Final Report on one phase of a continuing NRL Problem
		6. PERFORMING ORG. REPORT NUMBER
7. AUTHOR(s) G.V. Trunk		8. CONTRACT OR GRANT NUMBER(s)
9. PERFORMING ORGANIZATION NAME AND ADDRESS Naval Research Laboratory Washington, D.C. 20375		10. PROGRAM ELEMENT, PROJECT, TASK AREA & WORK UNIT NUMBERS NRL Problem R02-97 Program Element 61153N-21 Project RR021-05-41
11. CONTROLLING OFFICE NAME AND ADDRESS Department of the Navy Office of Naval Research Arlington, VA 22217		12. REPORT DATE July 15, 1977
		13. NUMBER OF PAGES 11
14. MONITORING AGENCY NAME & ADDRESS (if different from Controlling Office)		15. SECURITY CLASS. (of this report) Unclassified
		15a. DECLASSIFICATION/DOWNGRADING SCHEDULE
16. DISTRIBUTION STATEMENT (of this Report) Approved for public release; distribution unlimited		
17. DISTRIBUTION STATEMENT (of the abstract entered in Block 20, if different from Report)		
18. SUPPLEMENTARY NOTES		
19. KEY WORDS (Continue on reverse side if necessary and identify by block number) Radar MTI Integration Loss		
20. ABSTRACT (Continue on reverse side if necessary and identify by block number) MTI signal processing correlates the independent input noise and thus degrades detection performance when the MTI pulses are integrated. By use of simulation techniques the noise integration loss is calculated for small and intermediate sample sizes, for which the output is not Gaussian distributed. The losses for two-, three-, four-, and five-pulse MTIs are approximately 1.0, 1.8, 2.2, and 2.5 dB respectively.		



## MTI NOISE INTEGRATION LOSS

### INTRODUCTION

MTI signal processing correlates the receiver noise and thus results in degraded detection performance when the MTI pulses are integrated. Previous investigators [1,2] have described the decreased performance in terms of a reduction in the effective number of independent pulses integrated. However, since the effective number of pulses  $N_e$  can be represented by

$$N_e = \frac{(\sigma^2/m^2)_{in}}{(\sigma^2/m^2)_{out}},$$

where  $\sigma$  and  $m$  are the standard deviation and mean of the input samples,  $N_e$  has a precise meaning (in terms of detection performance) only if the output noise distribution is completely specified by  $N_e$ . For instance, when the number of pulses integrated ( $N$ ) is large, the integrated output is approximately Gaussian distributed and integration improvement varies as the square root of the number of pulses integrated. Thus the loss (due to the MTI correlating the receiver noise) in signal-to-noise ratio ( $S/N$ ) for a large number of integrated pulses is

$$L = 10 \log (N/N_e)^{1/2}.$$

In this report the MTI integration loss is calculated when the number of integrated pulses is small and thus the output is not Gaussian distributed. This calculation is performed using simulation techniques. First, the appropriate thresholds for a given probability of false alarm  $P_{fa}$  are calculated using importance-sampling techniques. Next, probability of detection  $P_D$  curves are generated by simulation of the pulse-by-pulse video. Finally, the MTI integration loss is found by comparing the generated  $P_D$  curves with those for independent samples [3].

### FALSE-ALARM THRESHOLDS

Although Monte-Carlo simulations have been used for many years to calculate  $P_D$  curves, they have not been used to calculate  $P_{fa}$  curves because of the enormous number of repetitions usually required: approximately  $10/P_{fa}$ . However this difficulty can be overcome by using importance sampling [4]. The fundamental principle of the importance-sampling technique is to modify the probabilities that govern the outcome of the basic experiment of the simulation in such a way that the event of interest (the false alarm) occurs more frequently. This distortion is then compensated for by weighting each event by the ratio of the probability that this specific event would have occurred if the true probabilities had been used in the simulation to the probability that this same event would occur with the distorted probabilities. Consequently by proper choice of the distorted

probabilities the number of repetitions can be reduced greatly. For instance, the mean of a function  $Q(x)$  is given by

$$E\{Q(x)\} = \int Q(x) dP(x),$$

where  $P(x)$  is the distribution of  $x$ . The mean of  $Q(x)$  can be estimated by selecting  $M$  independent samples  $x_i$  from  $P(x)$  and associating the probability  $1/M$  with each event. Then  $E\{Q(x)\}$  can be estimated by

$$\frac{1}{M} \sum_{i=1}^M Q(x_i). \quad (1)$$

The importance-sampling technique uses the Radon-Nikodym derivative to express the mean value of  $Q(x)$  by

$$E\{Q(x)\} = \int Q(x) \frac{dP(x)}{dG(x)} dG(x),$$

where  $G(x)$  is a distribution function. The mean  $E\{Q(x)\}$  can be estimated by selecting  $M$  independent samples from  $G(x)$  and associating the probability  $dP(x_i)/MdG(x_i)$  with each event  $Q(x_i)$ . Thus  $E\{Q(x)\}$  is estimated by

$$\frac{1}{M} \sum_{i=1}^M Q(x_i) \frac{dP(x_i)}{dG(x_i)}. \quad (2)$$

Since (1) and (2) are both unbiased estimates of  $Q(x)$ , it is possible to select  $G(x)$  so that the variance of (2) is less than the variance of (1).

In our problem of determining the threshold for a given  $P_{fa}$ , when MTI samples are noncoherently integrated, it is necessary to estimate the distribution curve

$$P(Z_j \leq T) = 1 - P_{fa}, \quad (3)$$

where

$$Z_j = \sum_{i=1}^N Z_{ij}, \quad (4)$$

in which

$$Z_{ij} = \left[ (x'_{ij})^2 + (y'_{ij})^2 / P(k) \right]^{1/2} \quad (5)$$

where, for a two-pulse MTI,

$$x'_{ij} = x_{ij} - x_{i-1,j} \quad (6)$$

and

$$y'_{ij} = y_{ij} - y_{i-1,j}, \quad (7)$$

with  $x_{ij}$  and  $y_{ij}$  being independent Gaussian variables with zero mean and a variance of  $\sigma$  and  $P(k)$  being the noise power out of a k-pulse MTI:  $P(2) = 2$ ,  $P(3) = 6$ ,  $P(4) = 20$ , and  $P(5) = 70$ . The straightforward way of estimating (7) is to generate Gaussian samples by

$$x_{ij} = \sigma(-2 \ln u_{ij})^{1/2} \sin 2\pi v_{ij} \quad (8)$$

and

$$y_{ij} = \sigma(-2 \ln u_{ij})^{1/2} \cos 2\pi v_{ij}, \quad (9)$$

with  $u_{ij}$  and  $v_{ij}$  being independent random numbers uniformly distributed on the interval (0,1). To estimate (3),  $M$  independent sums  $\{Z_j, j = 1, M\}$  are formed using (4) through (7), and the estimated distribution is

$$\hat{P}(Z \geq T) = \frac{1}{M} \sum_{j=1}^M \delta_j,$$

where

$$\begin{aligned} \delta_j &= 1, & Z_j &\geq T, \\ &= 0, & Z_j &< 0. \end{aligned}$$

Importance sampling differs from the previous procedure by generating samples using

$$x_{ij} = \alpha(-2 \ln u_{ij})^{1/2} \sin 2\pi v_{ij} \quad (10)$$

and

$$y_{ij} = \alpha(-2 \ln u_{ij})^{1/2} \cos 2\pi v_{ij}, \quad (11)$$

where  $\alpha > \sigma$ , a device which yields more false alarms. Using (10) and (11) and (4) through (7),  $M$  sums  $Z_j$  are generated. Then the estimated distribution is

$$\hat{P}(Z \geq T) = \frac{1}{M} \sum_{j=1}^M \delta_j P_j,$$

where

$$\begin{aligned}\delta_j &= 1, & Z_j &\geq T, \\ &= 0, & Z_j &< 0,\end{aligned}$$

and

$$P_j = \prod_{i=2-k}^N \frac{\frac{1}{2\pi\sigma^2} e^{-(x_{ij}^2 + y_{ij}^2)/2\sigma^2}}{\frac{1}{2\pi\alpha^2} e^{-(x_{ij}^2 + y_{ij}^2)/2\alpha^2}}.$$

With use of  $\alpha = 2.0$  and  $M = 20,000$  for  $N = 4$ ,  $\alpha = 1.7$  and  $M = 10,000$  for  $N = 8$ ,  $\alpha = 1.5$  and  $M = 10,000$  for  $N = 16$ , and  $\alpha = 1.3$  and  $M = 2500$  for  $N = 32$ , threshold curves were generated for two-, three-, four-, and five-pulse (binary weighting) MTIs and are shown in Fig. 1. The reference curve for independent samples was generated using detection curves in Robertson [3].

#### PROBABILITY OF DETECTION

Since the S/N out of the MTI is a function of the target doppler, the doppler frequency where the input and output S/N are equal will be used. The S/N gain (or loss) provided by the k-pulse MTI is

$$\frac{\left(\sum_{i=1}^k a_i \cos i\phi_k\right)^2 + \left(\sum_{i=1}^k a_i \sin i\phi_k\right)^2}{\sum_{i=1}^k a_i^2}, \quad (12)$$

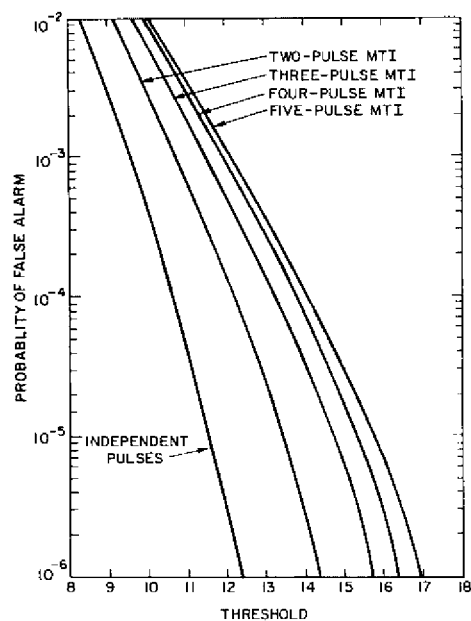
where  $\{a_i, i = 1, \dots, k\}$  are the MTI coefficients and  $\phi$  is the change in target phase between successive PRFs. Setting (12) equal to 1 and solving for  $\phi_k$  yields the solutions  $\phi_2 = 90^\circ$ ,  $\phi_3 = 103^\circ$ ,  $\phi_4 = 110.9^\circ$ , and  $\phi_5 = 116.5^\circ$ .

Thus the  $P_D$  for a k-pulse MTI and a given  $P_{fa}$  can be found by generating sample video using

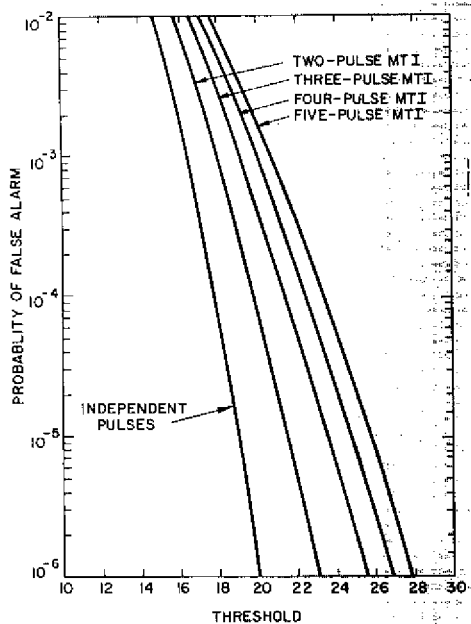
$$x_{ij} = \sigma(-2 \ln u_{ij})^{1/2} \sin 2\pi v_{ij} + A \sin i\phi_k \quad (13)$$

and

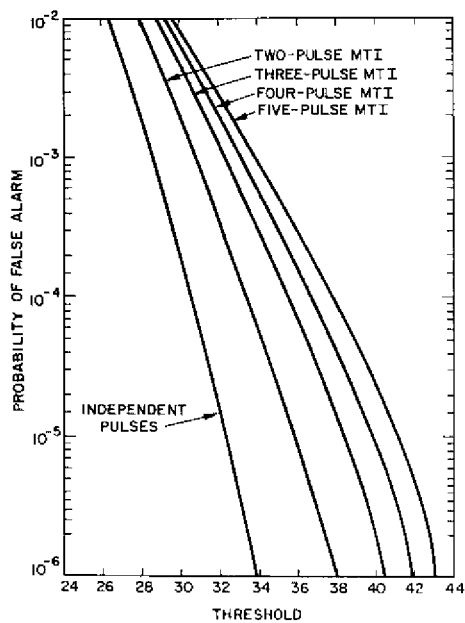
$$y_{ij} = \sigma(-2 \ln u_{ij})^{1/2} \cos 2\pi v_{ij} + A \cos i\phi_k, \quad (14)$$



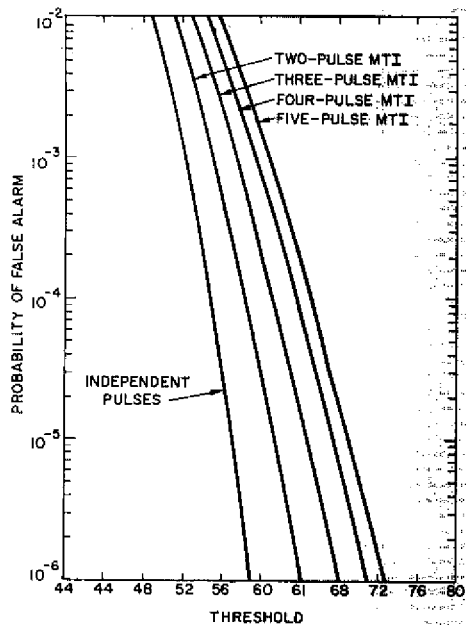
(a)  $N = 4$



(b)  $N = 8$



(c)  $N = 16$



(d)  $N = 32$

Fig. 1 — Threshold curves for  $N$  pulses integrated



where  $S/N(\text{dB}) = 10 \log (A^2/2\sigma^2)$ . By use of (13) and (14) and (3) through (7),  $M = 1024 Z_i$  values were generated for each  $S/N$  and compared to the appropriate threshold. The  $P_D$  curves for  $P_{fa} = 10^{-6}$  are shown in Fig. 2.

The difference between the  $P_D$  curves for the various MTIs and the curve for independent pulses is the MTI noise integration loss. This loss is given in Table 1 for the  $P_D$  and  $P_{fa}$  values indicated. The loss appears to be fairly independent of both  $N$ , the number of pulses integrated, and  $P_{fa}$ .

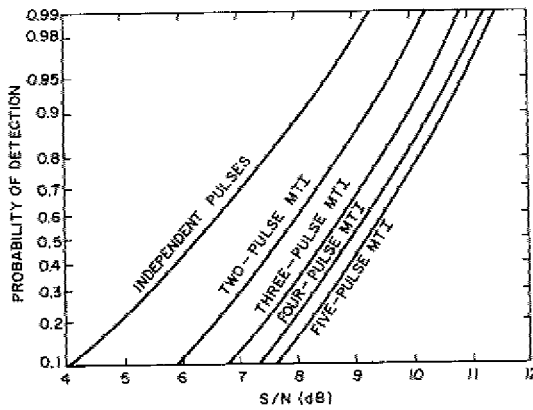
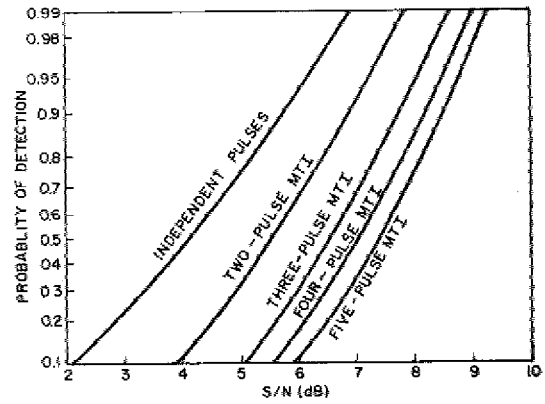
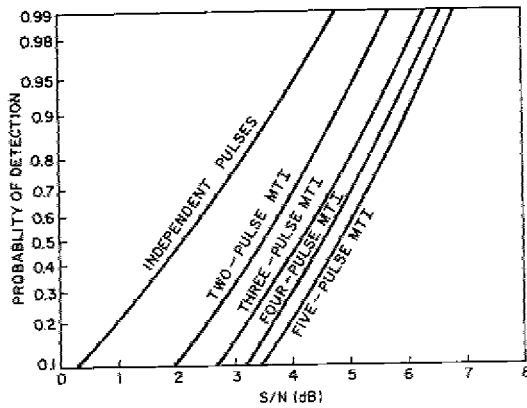
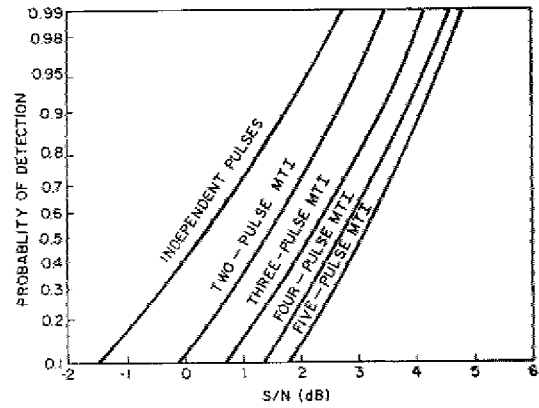
(a)  $N = 4$ (b)  $N = 8$ (c)  $N = 16$ (d)  $N = 32$ 

Fig. 2 — Probability of detection curves for  $N$  pulses integrated with  $P_{fa} = 10^{-6}$

Table 1 — MTI Noise Integration Loss for  $P_D = 0.9$   
and  $N$  Noncoherent Pulses Integrated

MTI Pulses	Loss (dB)				Average Difference (dB)
	$N = 4$	$N = 8$	$N = 16$	$N = 32$	
$P_{fa} = 10^{-6}$					
Two	1.1	1.1	1.1	0.9	1.0
Three	1.8	1.9	1.7	1.7	1.8
Four	2.2	2.4	2.1	2.1	2.2
Five	2.5	2.7	2.3	2.4	2.5
$P_{fa} = 10^{-4}$					
Two	1.1	0.9	0.9	0.8	0.9
Three	1.8	1.7	1.6	1.5	1.6
Four	2.1	2.1	1.9	1.9	2.0
Five	2.3	2.5	2.1	2.2	2.3

## COMPARISON WITH PREVIOUS RESULTS

The number of effective pulses integrated for a  $k$ -pulse MTI is given [1] by

$$N_e(k) = \frac{N^2}{N + 2 \sum_{j=1}^{N-1} (N-j)R_k^2(j)},$$

where  $R_k(j)$  is the correlation coefficient

$$R_k(j) = \frac{E\{x_i' x_{i+j}'\}}{P(k)}.$$

Thus, to find the MTI noise integration loss, the difference must be found between the required S/N for  $N_e$  and  $N$  independent pulses. To accomplish this, a curve of S/N versus  $N$  for  $P_D = 0.9$  and  $P_{fa} = 10^{-6}$  was generated using the detection curves in Robertson [3] and is shown in Fig. 3. From this curve the MTI noise integration loss was calculated and is shown in Table 2. These losses are about 0.2 dB higher than the corresponding losses in Table 1.

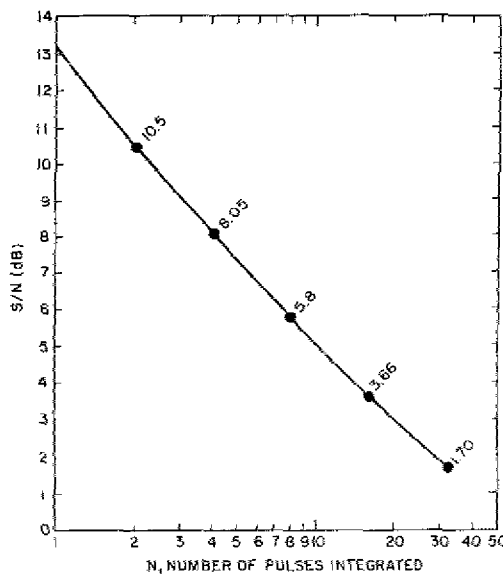


Fig. 3 —  $S/N$  for  $P_D = 0.9$  and  $P_{fa} = 10^{-6}$  as a function of the number of independent pulses integrated

Table 2 — MTI Noise Integration Loss Using the Effective Number of Pulses  $N_e$  Integrated for  $P_D = 0.9$  and  $P_{fa} = 10^{-6}$

MTI Pulses	Loss (dB)				Average Difference (dB)
	$N = 4$	$N = 8$	$N = 16$	$N = 32$	
Two	1.1	1.2	1.2	1.1	1.1
Three	1.8	1.7	1.9	1.8	1.8
Four	2.4	2.6	2.4	2.3	2.4
Five	2.7	2.9	2.9	2.7	2.8

## SUMMARY

MTI signal processing correlates the receiver noise, and this results in an MTI noise integration loss. The losses for two-, three-, four-, and five-pulse MTIs are approximately 1.0, 1.8, 2.2, and 2.5 dB respectively. The  $P_D$  for a given target can be found using the following procedure:

1. Calculate the input  $S/N$  (to the MTI) using the radar range equation;
2. Calculate the output  $S/N$  from the MTI using (12)
3. Use Fig. 2 to determine  $P_D$  or else assume all  $N$  pulses are independent, reduce  $S/N$  by the MTI noise integration loss, and find  $P_D$  from standard detection curves such as given in Robertson [3].

## REFERENCES

1. W.M. Hall and H.R. Ward, "Signal-to-Noise Loss in Moving Target Indicator," **IEEE Proceedings** 56 (No. 2), 233-234 (Feb. 1968).
2. F.F. Kretschmer, Jr., "Correlation Effects of MTI Filters," **IEEE Trans Aerospace and Electronic Systems** AES-13 (No. 3), 321-322 (May 1977).
3. G.H. Robertson, "Operating Characteristics For a Linear Detector of CW Signals in Narrow-Band Gaussian Noise," **Bell Sys. Tech. J.** 46, 755-774 (Apr. 1967).
4. V.G. Hansen, "Detection Performance of Some Nonparametric Rank Tests and an Application To Radar," **IEEE Trans. Inform. Theory** IT-16 (No. 3), 309-318 (May 1970).